

# EE 232: Lightwave Devices

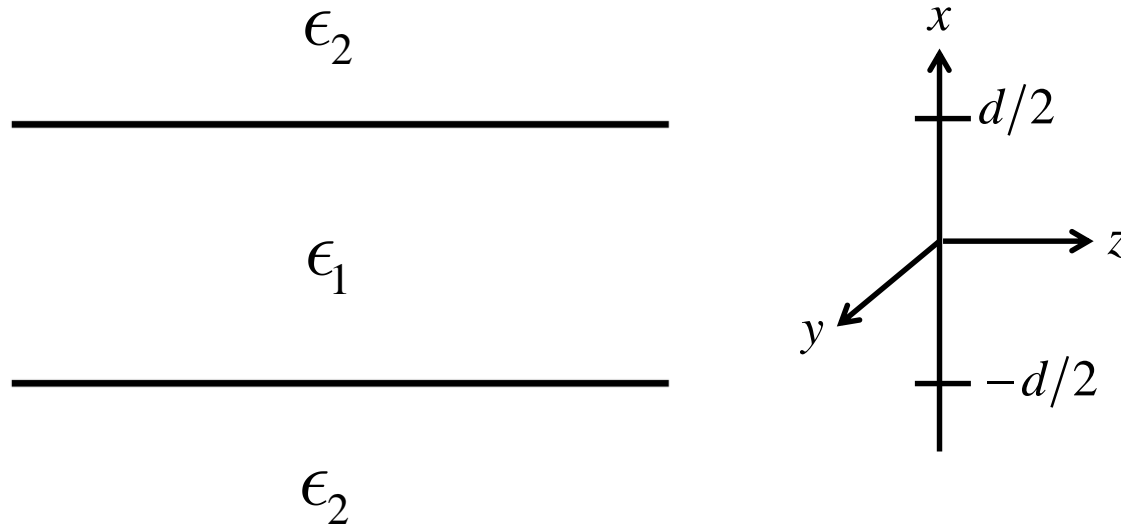
## Lecture #15 – Optical waveguides

**Instructor:** Seth A. Fortuna

Dept. of Electrical Engineering and Computer Sciences  
University of California, Berkeley

3/14/2019

# Slab waveguide



Slab waveguide consists of a slab of high-index material surrounded by low-index material ( $\epsilon_1 > \epsilon_2$ ). The waveguide is assumed to be infinitely large in the  $y$  and  $z$ -directions.

We wish to find confined electromagnetic modes that propagate in the  $+z$  direction and solve the source-free time-harmonic wave equation

$$\left(\nabla^2 + \omega^2 \mu \epsilon\right) \mathbf{E} = 0$$

# Wave equation

In general, solution for the electric-field can be written as

$$\mathbf{E}(x, y, z) = \hat{x}E_x(x, y, z) + \hat{y}E_y(x, y, z) + \hat{z}E_z(x, y, z)$$

If we plug back into the wave equation,

$$\nabla^2 \mathbf{E} + \omega^2 \mu \epsilon \mathbf{E} = \nabla^2 (\hat{x}E_x + \hat{y}E_y + \hat{z}E_z) + \omega^2 \mu \epsilon (\hat{x}E_x + \hat{y}E_y + \hat{z}E_z)$$

We separate into three equations

$$\nabla^2 E_x + \omega^2 \mu \epsilon E_x = 0 \quad (1)$$

$$\nabla^2 E_y + \omega^2 \mu \epsilon E_y = 0$$

$$\nabla^2 E_z + \omega^2 \mu \epsilon E_z = 0$$

Let's expand equation (1)

$$\nabla^2 E_x + \omega^2 \mu \epsilon E_x = \frac{\partial^2 E_x}{\partial y^2} + \frac{\partial^2 E_x}{\partial y^2} + \frac{\partial^2 E_x}{\partial z^2} + \beta^2 E_x = 0 \quad (2)$$

# Wave equation

Using separation of variables, we assume

$$E_x(x, y, z) = f(x)g(y)h(z)$$

If we plug back into equation (2)

$$\frac{1}{f} \frac{d^2 f}{dx^2} + \frac{1}{g} \frac{d^2 g}{dy^2} + \frac{1}{h} \frac{d^2 h}{dz^2} = -\beta^2$$

The sum of the terms can equal a constant only if each individual term is a constant

$$\therefore \frac{1}{f} \frac{d^2 f}{dx^2} = \beta_x^2 \quad \frac{1}{g} \frac{d^2 g}{dy^2} = \beta_y^2 \quad \frac{1}{h} \frac{d^2 h}{dz^2} = \beta_z^2$$

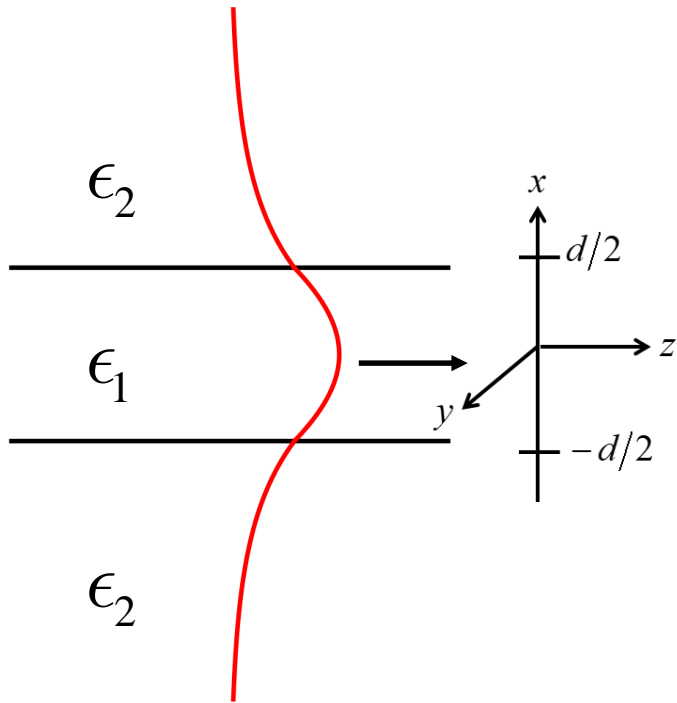
Typical solutions for these differential equations:

$$f(x) = A_1 e^{-i\beta_x x} + A_2 e^{i\beta_x x} \quad (\text{travelling wave})$$

$$f(x) = B_1 \cos(\beta_x x) + B_2 \sin(\beta_x x) \quad (\text{standing wave})$$

$$f(x) = C_1 e^{-i\beta_x x} \quad (\text{evanescent wave})$$

# Slab waveguide



We look for a solution that is transverse electric (TE), “bound” to the core of the waveguide, and travelling in the z-direction such that

$$\mathbf{E} = \hat{y}E_y(x, z) \rightarrow E_y(x, z) = f(x)h(z)$$

where we assume there is no dependence on y given the slab is translationally invariant in the y-direction.

Along the z-direction we expect a traveling wave solution

$$h(z) = C_1 e^{-i\beta_z z} + C_2 e^{i\beta_z z}$$

Along the x-direction, we expect a standing wave solution in the waveguide core and evanescent solution in the cladding.

$$f(x) = \begin{cases} A_1 e^{-\alpha(|x|-d/2)} & |x| > d/2 \\ A_2 \cos(\beta_x x) & -d/2 \leq x \leq d/2 \end{cases}$$

Even solution

$$f(x) = \begin{cases} B_1 e^{-\alpha(|x|-d/2)} & |x| > d/2 \\ B_2 \sin(\beta_x x) & -d/2 \leq x \leq d/2 \end{cases}$$

Odd solution

# Slab waveguide

① Plug into wave equation  $\longrightarrow$

$$\beta_x^2 + \beta_z^2 = \omega^2 \mu_1 \epsilon_1$$
$$-\alpha^2 + \beta_z^2 = \omega^2 \mu_2 \epsilon_2$$

② Apply boundary conditions at interface between core and cladding.

Tangential component of electric and magnetic field are equal across interface.

$$E_{y,core} \Big|_{x=\pm \frac{d}{2}} = E_{y,clad} \Big|_{x=\pm \frac{d}{2}}$$

$$H_{z,core} \Big|_{x=\pm \frac{d}{2}} = H_{z,clad} \Big|_{x=\pm \frac{d}{2}}$$

$$\alpha = \frac{\mu_2}{\mu_1} \beta_x \tan \left( \beta_x \frac{d}{2} \right) \text{ (even)}$$

$$\alpha = -\frac{\mu_2}{\mu_1} \beta_x \cot \left( \beta_x \frac{d}{2} \right) \text{ (odd)}$$

# Slab waveguide

③ After rearranging  $\longrightarrow$

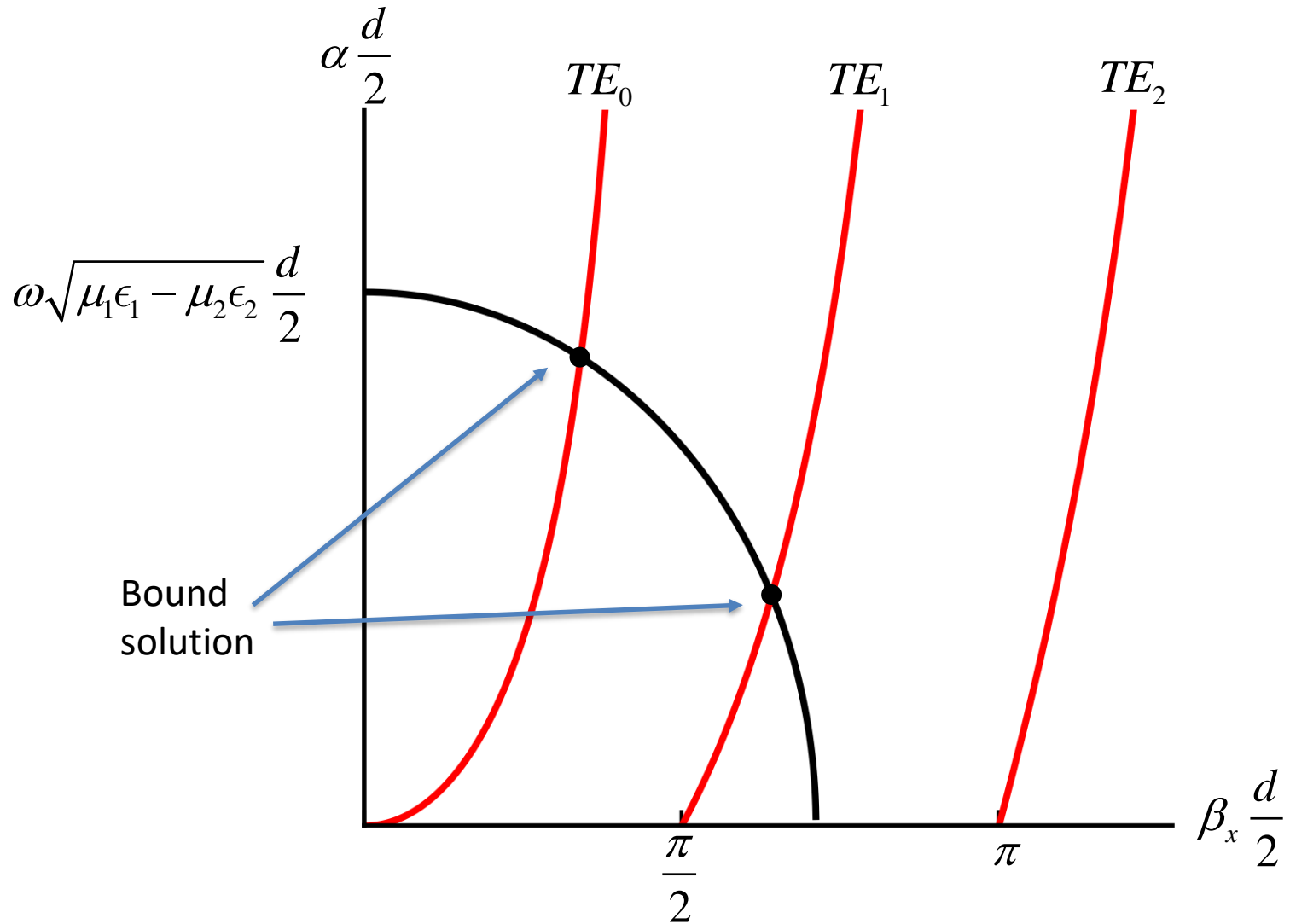
$$\left(\beta_x \frac{d}{2}\right)^2 + \left(\alpha \frac{d}{2}\right)^2 = \omega^2 (\mu_1 \epsilon_1 - \mu_2 \epsilon_2) \left(\frac{d}{2}\right)^2$$

$$\left(\alpha \frac{d}{2}\right) = \frac{\mu_2}{\mu_1} \left(\beta_x \frac{d}{2}\right) \tan\left(\beta_x \frac{d}{2}\right) \quad (\text{even})$$

$$\left(\alpha \frac{d}{2}\right) = -\frac{\mu_2}{\mu_1} \left(\beta_x \frac{d}{2}\right) \cot\left(\beta_x \frac{d}{2}\right) \quad (\text{odd})$$

# Slab waveguide

④ Solve graphically





# Cutoff condition

In the example on the previous slide we see that the  $TE_1$  mode would not have a solution and would be “cutoff” if the radius of the circle is less than  $\pi/2$

The cutoff condition for each mode can be generalized as

$$\omega(\mu_1\epsilon_1 - \mu_2\epsilon_2) \frac{d}{2} = m \frac{\pi}{2} \quad m=0,1,2,3\dots \quad (\text{Cutoff condition for } TE_m \text{ mode})$$

The waveguide will be single-mode if all modes except the fundamental mode are cutoff.

$$\omega(\mu_1\epsilon_1 - \mu_2\epsilon_2) \frac{d}{2} < \frac{\pi}{2} \quad (\text{Single mode condition})$$

# Effective index

**Effective index**  $n_{eff} = \frac{\beta_z}{\beta_0} \quad \beta_0 = \frac{2\pi}{\lambda_0}$

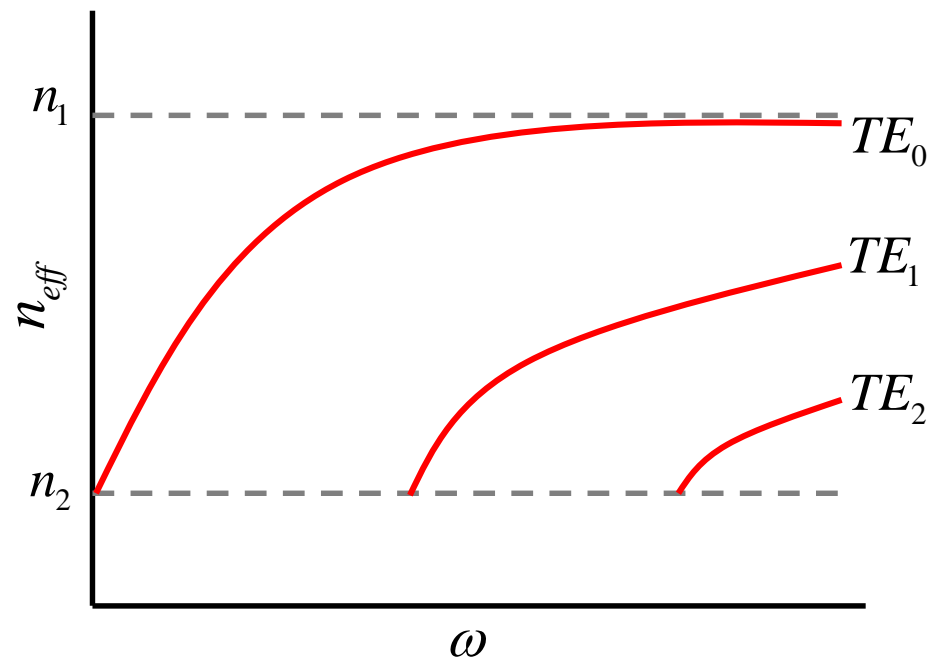
**High-frequency limit** Radius  $\rightarrow \infty$  as  $\omega \rightarrow \infty$   
 $\therefore \alpha \rightarrow \infty$

$$\begin{aligned} \beta_z^2 &= \omega^2 \mu_1 \epsilon_1 - \beta_x^2 \\ &\simeq \omega^2 \mu_1 \epsilon_1 \end{aligned} \quad n_{eff} = \frac{\beta_z}{\beta_0} = \frac{\sqrt{\mu_1 \epsilon_1}}{\sqrt{\mu_0 \epsilon_0}} = n_1 \quad \text{for } \mu_1 = \mu_0$$

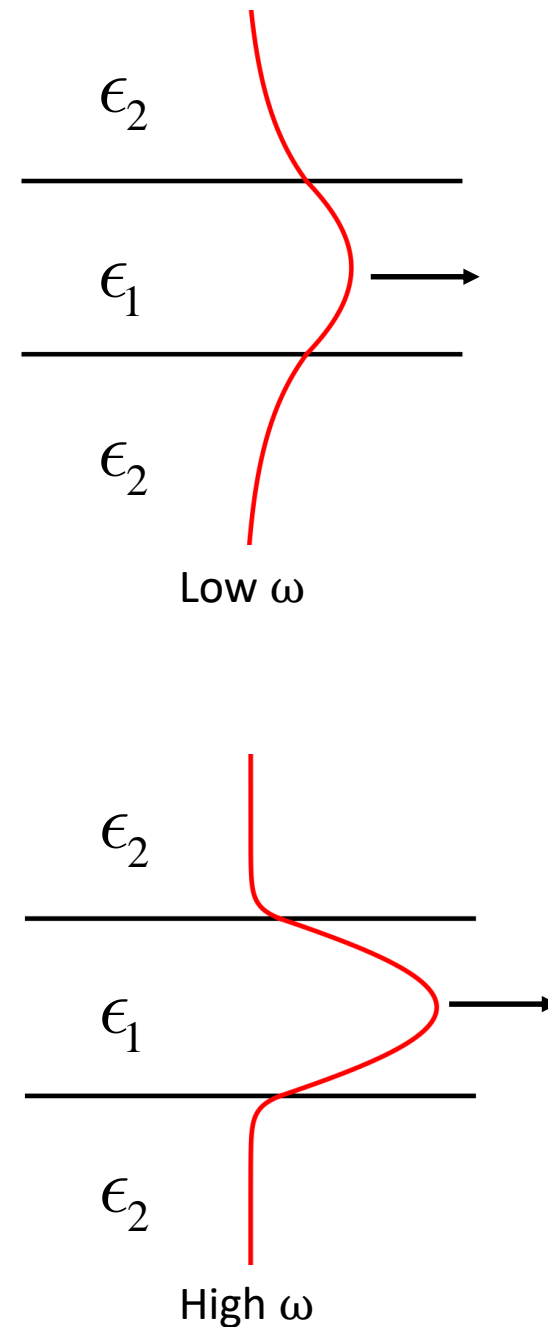
**Low-frequency limit** Radius  $\rightarrow 0$  as  $\omega \rightarrow 0$   
 $\therefore \alpha \rightarrow 0$

$$\begin{aligned} \beta_z^2 &= \omega^2 \mu_2 \epsilon_2 + \alpha^2 \\ &\simeq \omega^2 \mu_2 \epsilon_2 \end{aligned} \quad n_{eff} = \frac{\beta_z}{\beta_0} = \frac{\sqrt{\mu_2 \epsilon_2}}{\sqrt{\mu_0 \epsilon_0}} = n_2 \quad \text{for } \mu_2 = \mu_0$$

# Effective index



Effective index is a measure of how confined the mode is to the core



# Optical confinement factor

$$\Gamma = \frac{\text{Power in core}}{\text{Total power in mode}} = \frac{\frac{1}{2} \int_{core} \text{Re}(\mathbf{E} \times \mathbf{H}^*) \cdot \hat{z} dx}{\frac{1}{2} \int_{total} \text{Re}(\mathbf{E} \times \mathbf{H}^*) \cdot \hat{z} dx}$$

**Weak guidance limit (mode is mostly within cladding)**

$$\Gamma \simeq 2 \left( \frac{\pi d}{\lambda_0} \right)^2 (n_1^2 - n_2^2)$$

**For largest possible  $\Gamma$**

- (1) Thick core
- (2) Small wavelength
- (3) Large index contrast

# TM modes

$$\mathbf{H} = \hat{y}H_y(x, z) \rightarrow H_y(x, z) = f(x)h(z)$$

Along the z-direction we expect a traveling wave solution

$$h(z) = C_1 e^{-i\beta_z z} + C_2 e^{i\beta_z z}$$

Along the x-direction, we expect a standing wave solution in the waveguide core and evanescent solution in the cladding.

$$f(x) = \begin{cases} A_1 e^{-\alpha(|x|-d/2)} & |x| > d/2 \\ A_2 \cos(\beta_x x) & -d/2 \leq x \leq d/2 \end{cases}$$

Even solution

$$f(x) = \begin{cases} B_1 e^{-\alpha(|x|-d/2)} & |x| > d/2 \\ B_2 \sin(\beta_x x) & -d/2 \leq x \leq d/2 \end{cases}$$

Odd solution

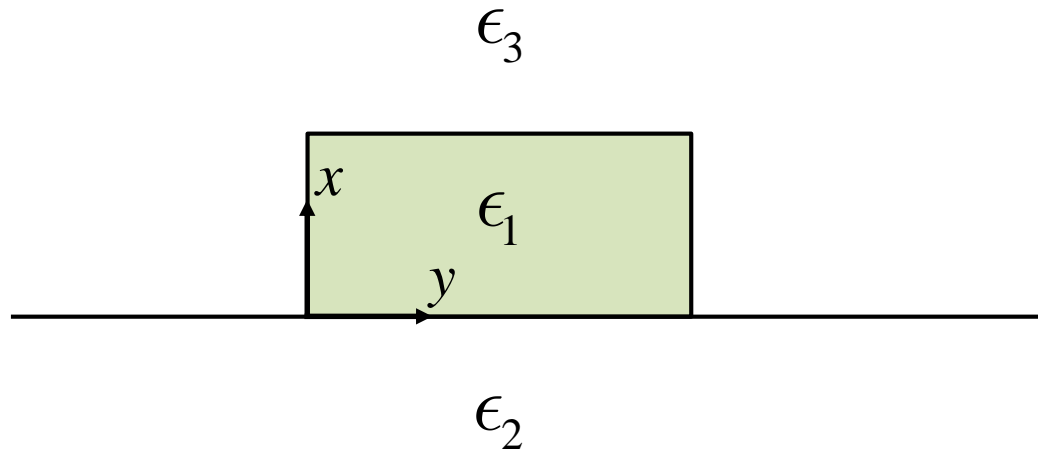
Eigenequations: 
$$\left(\beta_x \frac{d}{2}\right)^2 + \left(\alpha \frac{d}{2}\right)^2 = \omega^2 (\mu_2 \epsilon_2 - \mu_1 \epsilon_1) \left(\frac{d}{2}\right)^2$$

$$\left(\alpha \frac{d}{2}\right) = \frac{\epsilon_2}{\epsilon_1} \left(\beta_x \frac{d}{2}\right) \tan\left(\beta_x \frac{d}{2}\right)$$

$$\left(\alpha \frac{d}{2}\right) = -\frac{\epsilon_2}{\epsilon_1} \left(\beta_x \frac{d}{2}\right) \cot\left(\beta_x \frac{d}{2}\right)$$

# Rectangular waveguides

Rectangular waveguides have dielectric contrast in two-directions



Rectangular waveguides do not support pure TE or TM modes!  
Instead they support hybrid modes.

# Rectangular waveguides

## Hybrid modes

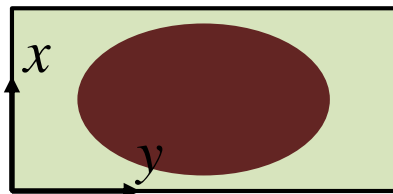
$HE_{pq}$   $H_x, E_y$  are the dominant components (quasi-TE)

$EH_{pq}$   $E_x, H_y$  are the dominant components (quasi-TM)

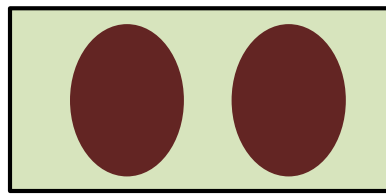
## Intensity patterns

$p \rightarrow$  number of nodes in the x-direction

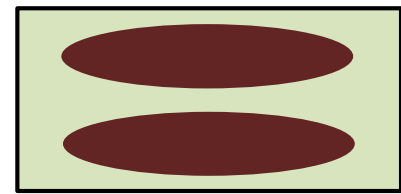
$q \rightarrow$  number of nodes in the y-direction



$HE_{00}$  or  $EH_{00}$

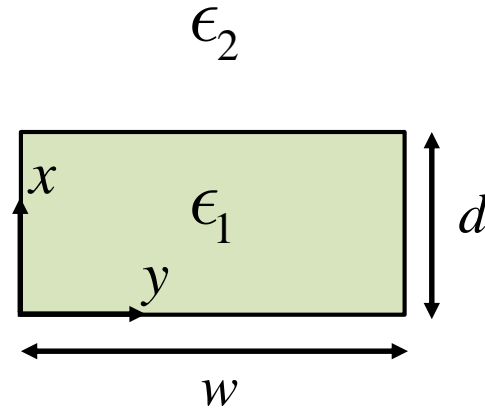


$HE_{01}$  or  $EH_{01}$



$HE_{10}$  or  $EH_{10}$

# Effective index method

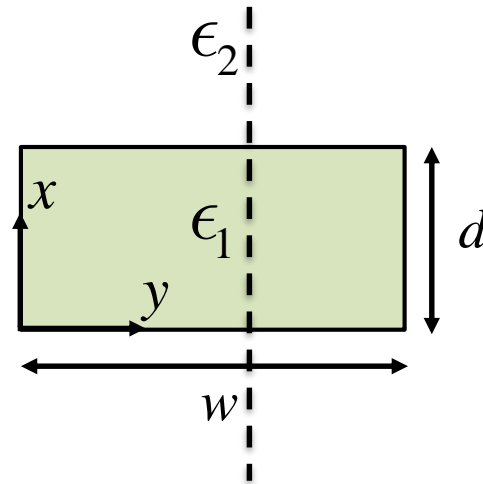


We estimate the propagation constant of the  $HE_{00}$  mode with the effective index method. We essentially break the 2D problem into a 1D slab waveguide problem.

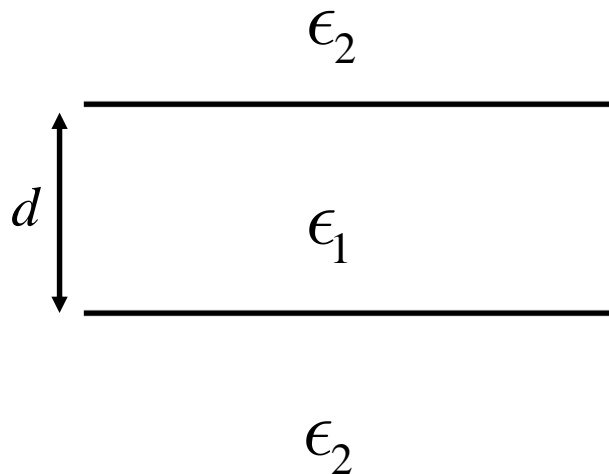
To simplify this problem we assume that the waveguide is completely surrounded by the same index. More sophisticated examples are found in the book.



# Step 1

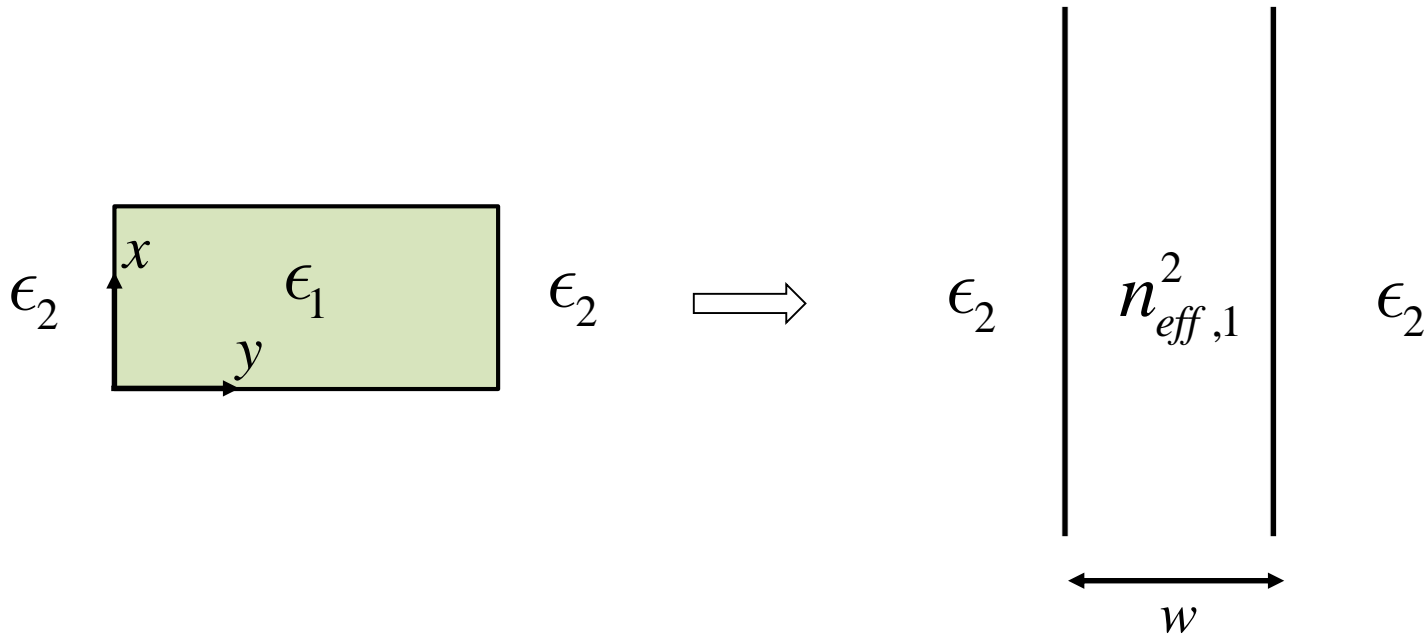


Solve for the **TE mode** of the slab waveguide with core of permittivity  $\epsilon_1$  and cladding with permittivity  $\epsilon_2$



Calculate the effective index  $n_{eff,1}$  and modal distribution  $F(x)$

# Step 2



Solve for the **TM mode** slab waveguide with core of permittivity  $n_{eff}^2$  and cladding with permittivity  $\epsilon_2$ . Calculate the propagation constant  $\beta_z$  and modal distribution  $G(y)$ .

The overall propagation constant of the 2D waveguide is then  $\beta_z$  and the modal distribution of the 2D waveguide is given by

$$E_y(x, y) = F(x)G(y)$$